

PROGRAMMABLE OSCILLATOR

PRIORITY CLAIM

[1] This application claims priority from Italian patent application No. MI2002A 002428, filed November 15, 2002, which is incorporated herein by reference.

TECHNICAL FIELD

[2] The present invention refers generally to a programmable oscillator and particularly to a programmable oscillator of the RC (Resistance, Capacitor) type. The present invention refers also to a controller circuit for lamps of the ballast type, and to an integrated circuit comprising a programmable oscillator of the RC type.

BACKGROUND

[3] An oscillator in some cases has to satisfy different requirements as for instance the following:

- allowing the oscillation frequency to vary with continuity in a prefixed range of frequencies;
- making available externally a terminal on which to act for varying its frequency;
- guaranteeing an excellent precision of the planned frequency.

[4] Besides, the possibility to integrate and to manage a great number of functions using a silicon area as small as possible has become a more and more important characteristic in a wide range of integrated circuits that comprise among the others the control devices for lamps of the ballast type, for power supplies, particularly those switching, for motors and many others.

[5] These integrated circuits often have an extremely reduced number of terminals relative to the number of functions that they perform.

[6] In view of the state of the art described, an embodiment of the present invention provides a programmable oscillator of the RC (Resistance, Capacitor) type that satisfies the above-listed requirements.

[7] According to this embodiment of the present invention, a programmable oscillator comprises a capacitor; a current generator couplable to said capacitor that generates a charging current of said capacitor; at least one resistance coupled to said capacitor; a comparator coupled to said capacitor for comparing a voltage at the terminals of said capacitor with a prefixed reference voltage and for generating an output signal; and a first switch, controlled by said output signal, coupled to said capacitor, that creates a current path to facilitate the discharging of said capacitor.

[8] Such a programmable oscillator can be included in a controller circuit for lamps of the ballast type comprising a half bridge that drives a lamp, wherein said half bridge is controlled by the oscillator.

[9] Furthermore, an integrated circuit may comprise such a programmable oscillator, where the integrated circuit comprises only a first and a second control pin arranged outside said integrated circuit; said current generator is coupled to said first pin; and said capacitor is coupled to said second pin.

[10] Consequently it is possible to implement a programmable oscillator of the RC type that has only two outside terminals and that also maintains high level precision and programmability.

BRIEF DESCRIPTION OF THE DRAWINGS

[11] The features and the advantages of the present invention will be made more evident by the following detailed description of a particular embodiment, illustrated as a non-limiting example in the annexed drawings, wherein:

[12] **FIG. 1** shows an exemplifying scheme of a controller circuit for lamps of the ballast type according to an embodiment of the invention;

[13] **FIG. 2** shows an exemplifying scheme of a programmable oscillator of the RC type, according to the prior art, used in a controller circuit for lamps of the ballast type;

[14] **FIG. 3** shows an exemplifying scheme of a management circuit of the programmable oscillator of the RC type according to the prior art of **FIG. 2**;

[15] FIG. 4 shows an exemplifying scheme of a programmable oscillator of the RC type according to an embodiment of the present invention;

[16] FIG. 5 shows the voltages available at some points of the circuit of FIG. 4 according to an embodiment of the invention.

5 [17] FIG. 6 shows a scheme of a programmable oscillator of the RC type according to another embodiment of the present invention.

[18] FIG. 7 shows the voltages available at some points of the circuit of FIG. 6 according to an embodiment of the invention.

DETAILED DESCRIPTION

10 [19] Referring now to FIG. 1, that shows an exemplifying scheme of a controller circuit for lamp of the ballast type according to an embodiment of the invention. A half bridge SP is constituted by transistors M1 and M2 and by respective diodes D1 and D2 is supplied by a substantially direct voltage Vcc provided by a pre-regulator stage PFC (Power Factor Corrector) placed upstream
15 and not shown.

[20] By means of a resistance RS connected in series to the half bridge SP, information is sent to a controller CONT, both with respect to the peak current of the half bridge SP, and with respect to the power of the lamp LAMP, by means of a filter RC composed by the resistance R and by the capacitor C, that filter the signal
20 coming from the resistance RS.

[21] The central point of the half bridge SP is connected to an inductance L1, therefore to the lamp LAMP and to the capacitor CB; a capacitor CH is connected in series with the filaments of the lamp LAMP.

25 [22] The capacitor CB, also called half-battery capacitor, charges itself at a voltage that is half that of the half bridge SP, and it has a sufficiently high value as to neglect the ripple voltage at its terminals.

[23] The capacitor CH has instead a fundamental role for the lighting of the lamp LAMP as, by resonating with the inductance L1, it allows the arc firing of the

lamp LAMP. However, when the lamp is turned, CH is short-circuited by the low impedance of the lamp LAMP.

[24] The transistors M1 and M2 are controlled by the controller CONT with the signals LS and HS respectively having a 50% operating cycle.

5 [25] The signals LS and HS are generated inside the controller CONT by the use of an RC oscillator operating by charge/discharge of a capacitor.

[26] We refer now to **FIG. 2** that shows an exemplifying scheme of a programmable oscillator of the RC type according to the prior art.

10 [27] It deals with a classical RC oscillator that can belong to an integrated circuit having three external terminals. A first terminal CF to which a terminal of the capacitor CCF is applied, the other terminal of the capacitor CCF is connected to ground. A second terminal RDT to which a terminal of a resistance RRDT is connected, the other terminal of the resistance RRDT is connected to ground. A third terminal RFMIN to which a terminal of a resistance RRFMIN and a terminal of a
15 resistance RF are connected, the other terminal of the resistance RRFMIN is connected to ground, the other terminal of the resistance RF is connected to the voltage V3.

[28] To the resistance RRDT is connected a voltage generator V1. It is composed by an operational amplifier OP1 to whose output the gate of a transistor
20 M3 is connected, whose source is connected to the resistance RRDT and to an input of the operational amplifier OP1, to the other input of the operational amplifier OP1 a prefixed voltage V4 is applied. The drain of the transistor M3 is connected to a current mirror connected to the voltage Vcc, formed by the transistor T1 connected as a diode and by the transistor T2 which is connected in turn to another current
25 mirror connected to ground, formed by the transistor T3 connected as a diode and by the transistor T4, whose collector is connected to a terminal A of a switch COM.

[29] To the resistance RRFMIN a voltage generator V2 is connected. It is composed by an operational amplifier OP2 to whose output the gate of a transistor
30 M4 is connected, whose source is connected to the resistance RRFMIN and to an input of the operational amplifier OP2, to the other input of the operational amplifier

OP2 a prefixed voltage V5 is applied. The drain of the transistor M4 is connected to a current mirror connected to the voltage Vcc, formed by the transistor T5 connected as a diode and by the transistor T6, whose collector is connected to a terminal B of a switch COM.

5 **[30]** The common terminal S of the switch COM is connected to the capacitor CCF.

[31] The signal available on the terminal CF is input to a management circuit CGO of the oscillator that outputs a control signal SC for the switch COM and the driving signals HS and LS for the half bridge SP.

10 **[32]** Two programmable currents are independently used to separately define both the charge and the discharge time. A first current generator is formed by the voltage generator V1 and by the resistance RRDT. A second current generator is formed by the voltage generator V2 and by the resistance RRFMIN.

15 **[33]** The discharging current of the capacitor CCF is defined by the resistance RRDT to which the fixed voltage V4 is applied. The charging current of the capacitor CCF is defined by the resistance RRFMIN to which the fixed voltage V5 is applied. The voltage V3 is used for varying the charging current of the capacitor CCF.

20 **[34]** Relative to the lamp-controller circuit of the ballast type, the discharging time defines the duration of the dead time, while the period of the oscillator is defined by the sum of the charging and discharging time.

25 **[35]** In **FIG. 3**, which is a block diagram of the management circuit CGO of the oscillator of **FIG. 2**, the voltage on the capacitor CCF, available on the terminal CF, is monitored by the window comparators COMP1 and COMP2 to which the reference voltages V6 and V7 are respectively applied. The outputs of the window comparators COMP1 and COMP2 are respectively applied to the set S and reset R inputs of a flip flop FF1. The outputs Q and NQ are applied to a control circuit CO that provides the control signal SC of the switch COM and therefore it controls the charging and the discharging of the capacitor CCF. Control circuit CO also provides
30 the driving signals HS and LS of the half bridge SP.

[36] Control circuit CO carries out also the following functions. It divides by two the frequency generated by the oscillator to control the half bridge SP with an operating cycle of 50%. It inserts at the half and at the end of the period the necessary dead times to avoid the cross conduction of the transistors M1 and M2 of **FIG. 1** (the dead times are timed by the discharge of the capacitor CCF). It eliminates the first cycle of the oscillator so that the half bridge SP starts the switching only from the second cycle of the capacitor CCF, since in the normal operation the oscillation occurs between two thresholds both different from zero, and the first cycle is longer since the capacitor CCF is initially discharged. It assures that at the lighting the transistor M2 is always turned on as first in order to be able to charge the capacitor CCF.

[37] We refer now to **FIG. 4** that shows an exemplifying scheme of a programmable oscillator of the RC type according to an embodiment of the present invention.

[38] It deals with an RC oscillator that can belong to an integrated circuit having only two external terminals. A first terminal TC to which a terminal of the capacitor C is applied, the other terminal of the capacitor C is connected to ground, in parallel to the capacitor C a resistance R1 is connected. A second terminal TR to which a terminal of a resistance R2 and a terminal of a resistance R3 are connected, the other terminal of the resistance R2 is connected to ground, the other terminal of the resistance R3 is connected to the voltage V10.

[39] To the terminal TR a voltage generator V11 is connected. It is formed by an operational amplifier OP3 to whose output the gate of a transistor M5 is connected, whose source is connected to the terminal TR and to an input of the operational amplifier OP3, to the other input of the operational amplifier OP3 is applied to a prefixed voltage V8. The drain of the transistor M5 is connected to a current mirror connected to the voltage Vcc, formed by the transistor T7 connected as a diode and by the transistor T8 which is connected in turn to a terminal A of a switch COM.

5 [40] The terminal B of the switch COM is not connected. The common terminal S of the switch COM is connected to the terminal TC. To the terminal TC an input of an operational amplifier OP4 is connected, to the other input of the operational amplifier OP4 the reference voltage V9 is applied. The output of the operational amplifier OP4 is connected to the gate of a transistor M6, whose source is connected to ground and its drain is connected to the terminal TC.

[41] The switch COM is controlled by the signal SC in a traditional way.

10 [42] The signal available on the terminal TC is input to a management circuit CGO of the oscillator that outputs the control signal SC to the switch COM and outputs the driving signals HS and LS of the half bridge SP (FIG. 1).

[43] The charging current of the capacitor C is determined by a current generator composed by the voltage generator V11 and by the resistance R2. The resistance R3 and the reference voltage V10 allow variation of this current according to the requirements of the application.

15 [44] By switching the switch COM from the position A to the position B the discharge of the capacitor C is obtained by means of the resistance R1 connected in parallel.

20 [45] When the voltage on the capacitor C reaches the reference voltage V9, the operational amplifier OP4 activates the transistor M6, which discharges the capacitor C by coupling it to ground via a low-resistance path.

[46] In FIG. 5 are shown the voltages available in some points of the circuit of FIG. 4 according to an embodiment of the invention.

25 [47] The voltage VC across the terminals of the capacitor C increases up to the reference voltage V6 (during the time Tc), then the switch COM switches from the position A to the position B, and the capacitor starts discharging down to the reference voltage V9 via the resistor R1 (during the time Ts). When the voltage VC across the capacitor C reaches V9, it discharges completely by means of the transistor M6.

[48] The driving signals LS and HS of the transistors M2 and M1 are also shown.

[49] T_{osc} denotes the oscillation period, and T_{sp} denotes the period of the half bridge SP.

5 [50] The programmable oscillator of FIG. 6 is similar to the programmable oscillator of FIG. 4; one difference is that in the oscillator of FIG. 6, one input of OP4 is coupled to the node B of the switch COM instead of to the terminal TC.

[51] The operation of the programmable oscillator of FIG. 6 is discussed in conjunction with FIGS. 3, 6, and 7.

10 [52] Referring to FIG. 7, during a first portion T_c of the oscillation period T_{osc} , the switch COM is in the A position such that the voltage generator V11 charges the capacitor C from a voltage V_L to the voltage V_6 . Then, the signal SC switches COM to the B position.

15 [53] During a second portion T_s of the oscillation period T_{osc} , the switch COM is in the B position. First, the capacitor C discharges through the resistor R1 from the voltage V_6 to the voltage V_9 (this discharge is represented by the exponentially decaying portions of the waveform in FIG. 7). When the voltage across the capacitor C is lower than V_9 , OP4 turns on the transistor M6, which discharges the capacitor C to the voltage V_L — in one embodiment, the voltage $V_L =$
20 0 volts, and in another embodiment, the voltage $V_L = V_7$, although V_L may have other values. Typically, the on resistance of the transistor M6 is much less than R_1 , such that the capacitor C discharges more quickly through M6 than through R_1 . Then, the signal SC switches COM back to the A position, and the oscillator period T_{osc} repeats.

25 [54] From the foregoing it will be appreciated that, although specific embodiments of the invention have been described herein for purposes of illustration, various modifications may be made without deviating from the spirit and scope of the invention.